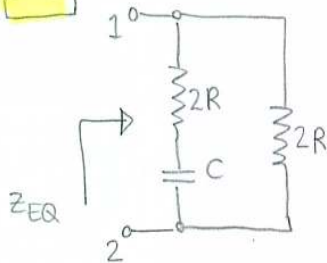


HW7 - MAE 140 fall 2012

10.3, 10.5, 10.9, 10.17, 10.19, 10.25, 10.27, 10.32, 10.61, 10.68

10.3



$$\begin{aligned} a) \quad Z_{EQ} &= \left(2R + \frac{1}{sC} \right) \parallel 2R = \left(\frac{1 + 2sCR}{sC} \right) \parallel 2R \\ &= \frac{(1 + 2sCR) 2R}{sC} = \frac{1 + 2sCR}{\frac{1}{2R} + sC + sC} \\ &= \frac{(1 + 2sCR) 2R}{sC} \end{aligned}$$

$$Z_{EQ} = \frac{sR + \frac{1}{2C}}{s + \frac{1}{4RC}}$$

zero: $s = -\frac{1}{2CR}$

pole: $s = -\frac{1}{4RC}$

b) Want pole at -640 rad/s

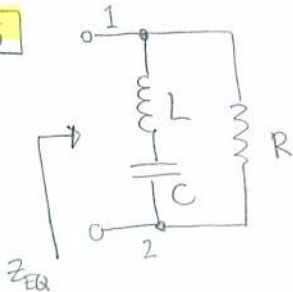
- select $R = 1 \text{ k}\Omega = 1000 \Omega$

$$-640 \text{ rad/s} = -\frac{1}{4RC}$$

$$C = \frac{1}{4R \cdot 640} = 3.9063 \cdot 10^{-7} = 0.391 \mu\text{F} = 391 \text{ nF}$$

$$-s_{\text{zero}} = -\frac{1}{2CR} = 2 \cdot s_{\text{pole}} = -1280 \text{ rad/s}$$

10.5



$$a) \quad z_{EQ} = \left(sL + \frac{1}{sC} \right) \parallel R = \left(\frac{1 + s^2 LC}{sC} \right) \parallel R$$

$$z_{EQ} = \frac{\left(\frac{1 + s^2 LC}{sC} \right) R}{\frac{1 + s^2 LC}{sC} + R} = \boxed{\frac{R + s^2 LCR}{1 + s^2 LC + sCR}} \quad z_{EQ}$$

$$z_{EQ} = \frac{s^2 R + \frac{R}{LC}}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

$$s_{1/2} = \pm j \sqrt{\frac{1}{LC}}$$

zeros

$$s_{1/2} = -\frac{R}{2L} \pm \sqrt{\left(\frac{R}{2L} \right)^2 - \frac{1}{LC}}$$

$$s_{1/2} = \frac{-RC \pm \sqrt{R^2 C^2 - 4LC}}{2LC}$$

poles

b) $R = 2 \text{ k}\Omega$
 $C = 0.1 \mu\text{F}$ } L such that zero at $s = \pm j 5000 \text{ rad/s}$?

$$(5000)^2 = \frac{1}{LC}$$

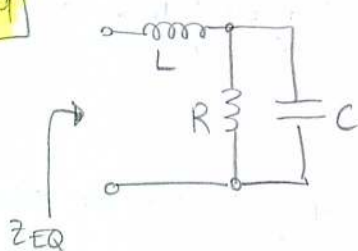
$$L = \frac{1}{C \cdot 5000^2} = \frac{1}{0.1 \cdot 10^{-6} \cdot 25 \cdot 10^6} = \frac{1}{2.5} = 0.4 \text{ H} = \boxed{400 \text{ mH}}$$

$$c) \quad s_{1/2} = -\frac{R}{2L} \pm \sqrt{\left(\frac{R}{2L} \right)^2 - \frac{1}{LC}}$$

$$s_{1/2} = -\frac{2 \cdot 10^3}{2 \cdot 0.4} \pm \sqrt{\frac{10^6}{0.16} - 5000^2}$$

$$\boxed{s_{1/2} = -2500 \pm j 4330.1}$$

10.9



$$L = 2 \text{ H}$$

$$R = 1 \text{ k}\Omega$$

$$C = 0.5 \mu\text{F}$$

$$a) \quad z_{EQ} = sL + R \parallel \frac{1}{sC}$$

$$= sL + \frac{R}{1 + sCR} = \frac{R + sL + s^2 LCR}{1 + sCR}$$

$$z_{EQ} = \frac{\frac{1}{C} + s \frac{L}{CR} + s^2 L}{s + \frac{1}{CR}}$$

$$\text{poles: } s = -\frac{1}{CR} = -\frac{1}{10^3 \cdot 0.5 \cdot 10^{-6}} = \boxed{-2000 \text{ rad/sec}} \quad \text{pole}$$

$$\text{zeros: } s^2 + s \frac{1}{CR} + \frac{1}{LC} = 0$$

$$s_{1/2} = -\frac{1}{2CR} \pm \sqrt{\left(\frac{1}{2CR}\right)^2 - \frac{1}{LC}}$$

$$s_{1/2} = -\frac{1}{2CR} = \boxed{-1000 \text{ rad/sec}}$$

2 zeros at -1000 rad/sec

$$b) \quad L = 5 \text{ H} \quad \bullet \text{ poles don't change: } s = -2000 \text{ rad/sec}$$

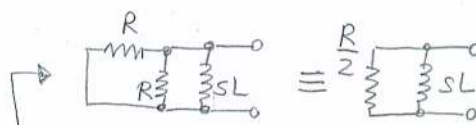
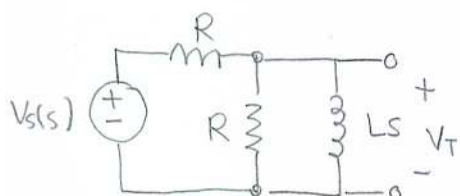
$\bullet$ zeros:

$$s_{1/2} = -1000 \pm \sqrt{10^6 - 0.4 \cdot 10^6} = -1000 \pm 774.6 \text{ rad/sec}$$

$$s_1 = -225.4 \text{ rad/sec}$$

$$s_2 = -1774.6 \text{ rad/sec}$$

10.17 Thevenin equivalent



$$1. \text{ For } Z_T \text{ set } V_s(s) = 0$$

$$Z_T = sL \parallel (R \parallel R) = sL \parallel \frac{R}{2} = \frac{sLR}{\frac{R}{2} + sL}$$

$$\boxed{Z_T = \frac{sLR}{R + 2sL}}$$

$$2. \text{ For } V_T \text{ open circuit}$$

$$V_T(s) = \frac{(R \parallel sL) V_s(s)}{R \parallel sL + R} = \frac{\left(\frac{sLR}{sL + R}\right) V_s(s)}{\frac{sLR}{sL + R} + R} = \frac{sLR}{sLR + sLR + R^2} V_s(s) = \boxed{\frac{sL}{2sL + R} V_s(s)}$$

$$3. \text{ Want pole at } s = -12 \text{ k rad/s:}$$

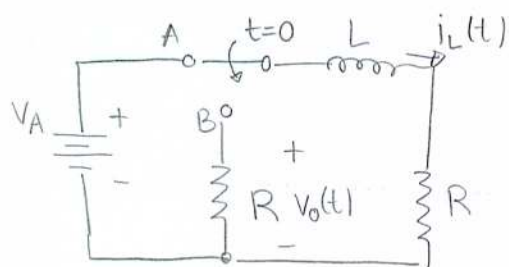
$$2sL + R = 0$$

$$s = -\frac{R}{2L} = -12 \text{ k} \quad \text{select } \boxed{R = 3 \text{ k}\Omega}$$

$$L = \frac{3 \text{ k}}{2 \cdot 12 \text{ k}} = \frac{1}{8} \text{ H} = \boxed{125 \text{ mH}}$$

NOTE: Students can select different values for R, resulting L will be $L = \frac{R}{24000}$

10.19 Solve for $I_L(s)$, $i_L(t)$, $V_o(s)$, $v_o(t)$; switch is moved to position B at time $t=0$

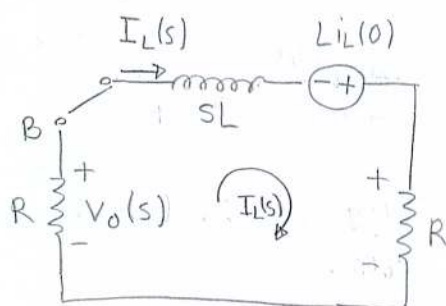


Initial condition

For $t < 0$ the circuit is in a DC steady-state condition and the inductor acts like a short circuit ($\text{inductor} \equiv \text{short}$).

Consequently, $i_L(t) = \frac{V_A}{R}$

First, transform the circuit into the s domain for $t > 0$



1. Using KVL we obtain

$$sL I_L(s) - L i_L(0) + R I_L(s) + R I_L(s) = 0$$

$$(sL + 2R) I_L(s) = L i_L(0) = L \frac{V_A}{R}$$

$$I_L(s) = \frac{L \frac{V_A}{R}}{sL + 2R} = \frac{\frac{1}{R}}{s + \frac{2R}{L}} V_A$$

2.

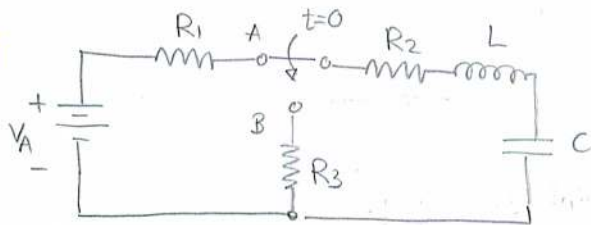
$$V_o(s) = -R I_L(s) = -\frac{V_A}{s + \frac{2R}{L}}$$

$$v_o(t) = \left\{ -V_A e^{-\frac{2R}{L}t} \right\} u(t)$$

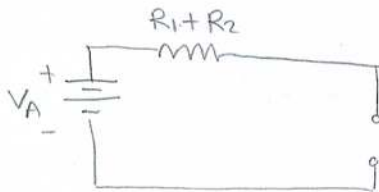
$$I_L(s) = \frac{\frac{1}{R}}{s + \frac{2R}{L}} V_A$$

$$i_L(t) = \left\{ \frac{V_A}{R} \cdot e^{-\frac{2R}{L}t} \right\} u(t)$$

10.25

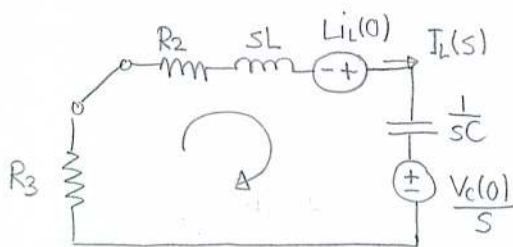


- a) First we have to find the initial inductor current and capacitor voltage. The circuit is in a DC steady-state condition for $t < 0$ and the inductor acts like a short circuit and the capacitor like an open circuit



Thus $i_L(0) = 0$
 $v_C(0) = V_A$

Now transform the circuit into the s-domain for $t > 0$



KVL:
 $(R_2 + R_3 + sL + \frac{1}{sC}) I_L(s) + \frac{V_A(0)}{s} = 0$

$$I_L(s) = - \frac{V_A}{s^2 L + (R_2 + R_3)s + \frac{1}{C}}$$

$$I_L(s) = - \frac{\frac{V_A}{L}}{s^2 + \frac{R_2 + R_3}{L}s + \frac{1}{LC}}$$

- b) $R_1 = R_2 = 500 \Omega$
 $R_3 = 1 k\Omega$
 $L = 0.5 \text{ H}$
 $C = 0.2 \mu\text{F}$
 $V_A = 15 \text{ V}$

Have to find $\mathcal{L}^{-1}\{I_L(s)\}$ first

$$I_L(s) = - \frac{30\text{V}}{s^2 + 3000s + 10 \cdot 10^6} \text{ A} = \frac{-30\text{V}}{(s^2 + 1500)^2 + 7.750.000} \text{ A}$$

NOTE: $\frac{\beta}{(s+\alpha)^2 + \beta^2} \xrightarrow{\mathcal{L}^{-1}} \{e^{-\alpha t} \sin \beta t\} u(t)$

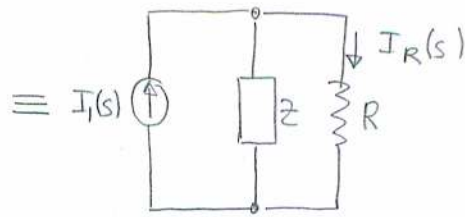
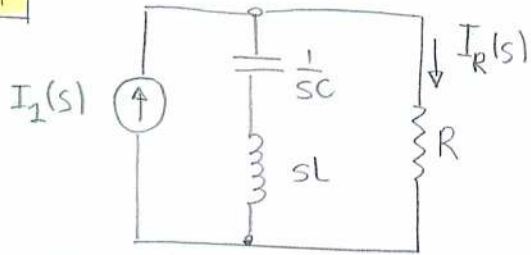
$$I_L(s) = \frac{-0.0108 \cdot 2784}{(s^2 + 1500^2)^2 + 2784^2} \text{ A}$$

$$i_L(t) = [-0.0108 \cdot e^{-1500t} \cdot \sin\{2784t\}] u(t) \text{ A}$$

Alternative way:

Poles at: $s_{1/2} = -1500 \pm j\sqrt{7.750.000} \rightarrow I_L(s) = \frac{k_1}{(s-s_1)} + \frac{k_2}{(s-s_2)} \rightarrow$ should obtain the same result

10.27



$$Z = \frac{1}{sC} + sL = \frac{1 + s^2 LC}{sC}$$

Current division

$$\frac{I_R(s)}{I_1(s)} = \frac{Z}{Z + R} = \frac{\frac{1 + s^2 LC}{sC}}{\frac{1 + s^2 LC}{sC} + R} = \boxed{\frac{1 + s^2 LC}{1 + sCR + s^2 LC}}$$

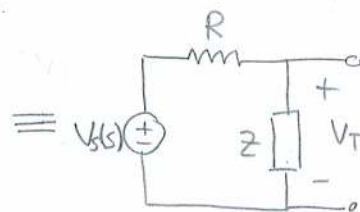
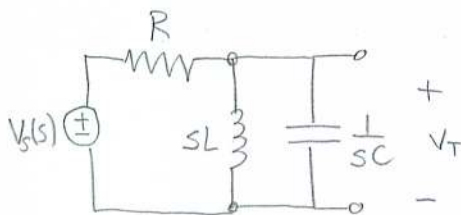
zeros: $1 + s^2 LC = 0$
 $s = \pm j\sqrt{\frac{1}{LC}}$

poles: $s^2 + \frac{R}{L}s + \frac{1}{LC} = 0$

$$s_{1/2} = -\frac{R}{2L} \pm \sqrt{\frac{R^2}{4L^2} - \frac{1}{LC}} = \frac{-RC \pm \sqrt{R^2 C^2 - 4LC}}{2LC}$$

10.31

Thevenin equivalent



$$Z = sL \parallel \frac{1}{sC} = \frac{\frac{sL}{sC}}{sL + \frac{1}{sC}} = \frac{sL}{1 + s^2 LC}$$

2. Equivalent resistance

set $V_s(s) = 0$

$$Z_{EQ} = Z \parallel R = \frac{SLR}{s^2 LCR + sL + R}$$

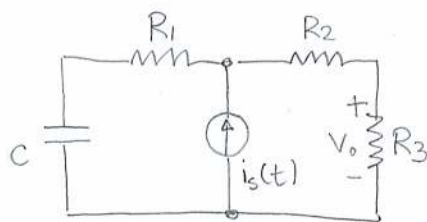
$$\boxed{Z_{EQ} = \frac{\frac{1}{C}s}{s^2 + \frac{1}{RC}s + \frac{1}{LC}}}$$

1. Thevenin voltage

$$V_T = \frac{Z}{R + Z} V_s(s) = \frac{sL}{R + sL + s^2 LCR} V_s(s)$$

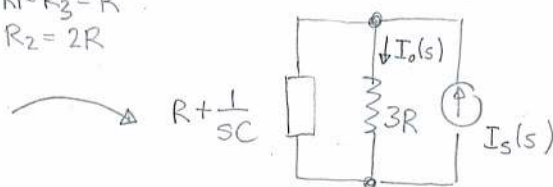
$$\boxed{V_T = \frac{\frac{1}{RC}s}{\frac{1}{LC} + \frac{1}{CR}s + s^2} V_s(s)}$$

10.32 Compute $v_o(t)$ via current division (no initial conditions)



$$R_1 = R_3 = R$$

$$R_2 = 2R$$



$$R_1 = R_3 = 5 \text{ k}\Omega$$

$$R_2 = 10 \text{ k}\Omega$$

$$C = 0.2 \mu\text{F} = 0.2$$

$$i_s(t) = 5 e^{-500t} u(t) \text{ mA}$$

$$I_s(t) = \frac{0.005}{s + 500} \text{ [A]}$$

Using element voltage-current relation

$$V_o(s) = R_3 I_o(s)$$

Calculate $I_o(s)$ using current division

$$I_o(s) = \frac{R + \frac{1}{sC}}{(R + \frac{1}{sC}) + 3R} I_s(s) = \frac{1 + sCR}{1 + 4sCR} I_o(s)$$

$$I_o(s) = \frac{1}{4} \cdot \frac{s + \frac{1}{CR}}{s + \frac{1}{4CR}} I_o(s)$$

$$V_o(s) = R_3 I_o(s) = \frac{R}{4} \cdot \frac{s + \frac{1}{CR}}{s + \frac{1}{4CR}} \cdot \frac{0.005}{s + 500}$$

NOTE:

$$\frac{1}{CR} = \frac{1}{0.2 \cdot 10^{-6} \cdot 5 \cdot 10^3} = 10^3$$

$$V_o(s) = \frac{25}{4} \cdot \frac{(s + 1000)}{(s + 250)(s + 500)} = \frac{k_1}{s + 250} + \frac{k_2}{s + 500}$$

$$k_1 = \lim_{s \rightarrow -250} \frac{25}{4} \cdot \frac{s + 1000}{s + 500} = \frac{25}{4} \cdot \frac{750}{250} = \frac{75}{4} = 18.75$$

$$k_2 = \lim_{s \rightarrow -500} \frac{25}{4} \cdot \frac{s + 1000}{s + 250} = \frac{25}{4} \cdot \frac{500}{-250} = -\frac{50}{4} = -12.5$$

$$V_o(s) = \frac{18.75}{s + 250} - \frac{12.5}{s + 500}$$

$$V_o(t) = \left\{ 18.75 e^{-250t} - 12.5 e^{-500t} \right\} u(t)$$

NOTE TO READER:

Students were supposed to solve 10.32; however, I scanned in problem 10.31;

Please give full credit if either 10.31 or 10.32 was solved correctly. Thanks!

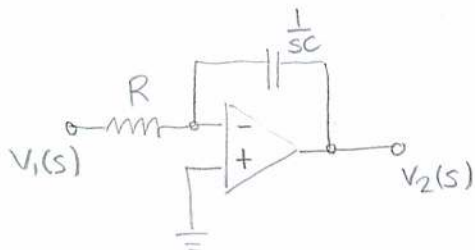
10.61

$$v_1(t) = \cos 1000t$$

$$\text{desired output } v_2(t) = -\sin 1000t$$

$$V_1(s) = \frac{s}{s^2 + 1000^2}$$

$$V_2(s) = -\frac{1000}{s^2 + 1000^2}$$



This is an inverting amplifier and we know that

$$V_2(s) = K V_1(s) = -\frac{\frac{1}{sC}}{R} V_1(s) = -\frac{1}{sCR} V_1(s)$$

$$\text{So } V_2(s) = -\frac{1}{sCR} V_1(s)$$

$$-\frac{1000}{s^2 + 1000^2} = -\frac{1}{sCR} \cdot \frac{s}{s^2 + 1000^2}$$

$$\rightarrow \text{Comparison shows that } 1000 = \frac{1}{CR}$$

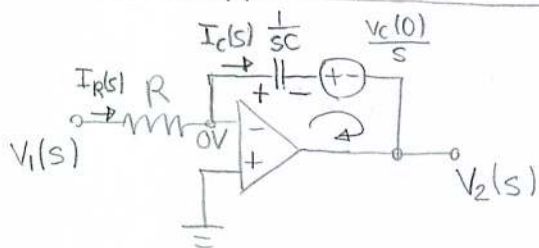
$$C = \frac{1}{1000R}$$

$$C = 1 \mu\text{F}$$

$$\text{select } R = 1 \text{ k}\Omega$$

NOTE: Students can choose values for R or C randomly and the other value can be computed via $C = \frac{1}{1000R}$

What happens if $v_c(0) = 10 \text{ V}$?



We know that $i_N = i_P = 0$ so $I_C(s) = I_R(s) = I(s)$
 $V^+ = V^- = 0$

$$\text{KVL: } -V_2(s) = \frac{1}{sC} I(s) + \frac{V_c(0)}{s} \quad I(s) = \frac{V_1(s)}{R}$$

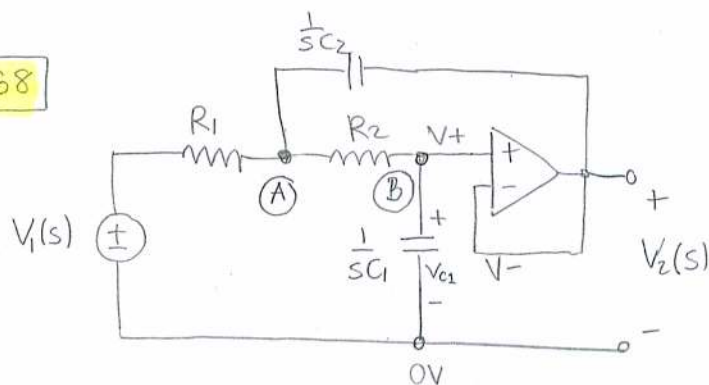
$$-V_2(s) = \frac{1}{sC} \cdot \frac{V_1(s)}{R} + \frac{V_c(0)}{s}$$

$$V_2(s) = -\frac{V_1(s)}{sCR} - \frac{V_c(0)}{s}$$

= step function in time domain

→ output is shifted by $V_c(0) = 10 \text{ V}$

10.68



Note: $i_N = i_P = 0$ and $V^+ = V^- = V_2(s)$

Now use KCL at node (A) and (B) to find $V_2(s)$

$$\begin{aligned} \text{KCL @ (A)} \quad \frac{V_1(s) - V_A(s)}{R_1} + \frac{V_2(s) - V_A(s)}{\frac{1}{sC_2}} + \frac{0V - V_A(s)}{R_2 + \frac{1}{sC_1}} &= 0 \\ \frac{1}{R_1} V_1(s) + sC_2 V_2(s) &= \left(\frac{1}{R_1} + sC_2 + \frac{sC_1}{1 + sC_1 R_2} \right) V_A(s) \end{aligned} \quad (1)$$

$$\begin{aligned} \text{KCL @ (B)} \quad \frac{V^+(s) - V_A(s)}{R_2} + \frac{V^+(s) - 0V}{\frac{1}{sC_1}} &= 0 \quad V^+(s) = V^-(s) = V_2(s) \\ V_2(s) - V_A(s) + sC_1 R_2 V_2(s) &= 0 \end{aligned}$$

$$V_A = (1 + sC_1 R_2) V_2(s) \quad (2)$$

Substitute (2) in (1)

$$\begin{aligned} \frac{1}{R_1} V_1(s) + sC_2 V_2(s) &= \left(\frac{1 + sC_2 R_1}{R_1} + \frac{sC_1}{1 + sC_1 R_2} \right) (1 + sC_1 R_2) V_2(s) \\ \frac{1}{R_1} V_1(s) &= \left\{ \frac{1}{R_1} (1 + sC_2 R_1)(1 + sC_1 R_2) + sC_1 - sC_2 \right\} V_2(s) \end{aligned}$$

$$\frac{V_1(s)}{V_2(s)} = 1 + s(C_1 R_2 + C_2 R_1) + s^2 C_1 C_2 R_1 R_2 + sC_1 R_1 - sC_2 R_1$$

$$V_2(s) = \frac{1}{s^2 C_1 C_2 R_1 R_2 + s(C_1 R_2 + C_1 R_1) + 1} V_1(s)$$

$$V_2(s) = \frac{\frac{1}{C_1 C_2 R_1 R_2}}{s^2 + s\left(\frac{1}{C_2 R_1} + \frac{1}{C_2 R_2}\right) + \frac{1}{C_1 C_2 R_1 R_2}} V_1(s)$$