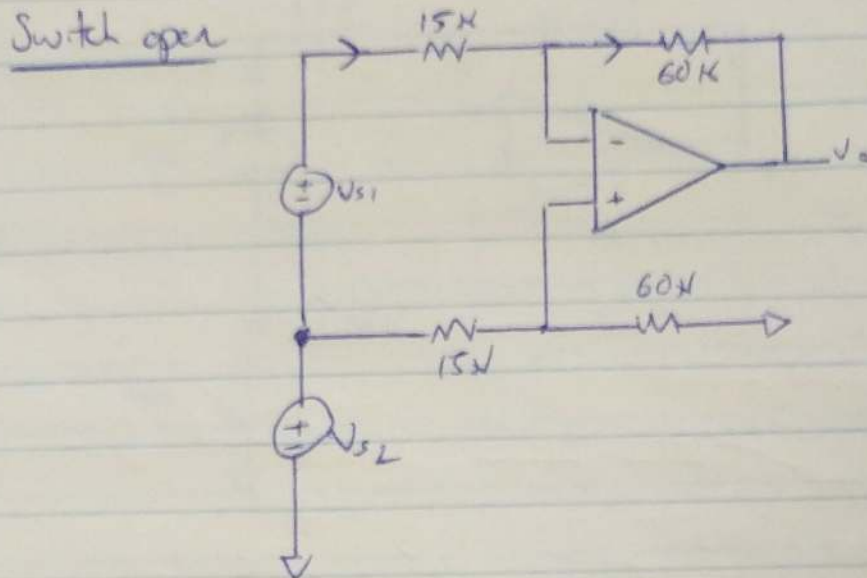


4.44



First, let us compute the voltage at the positive terminal

$$V_+ = \frac{V_{s2} \cdot 60k}{75k} = \frac{4V_{s2}}{5}$$

Now, we know that $V_+ = V_-$, so after assigning current directions we have that

$$\frac{V_{s1} + V_{s2} - \frac{4V_{s2}}{5}}{15k} = \frac{\frac{4V_{s2}}{5} - V_o}{60k}$$

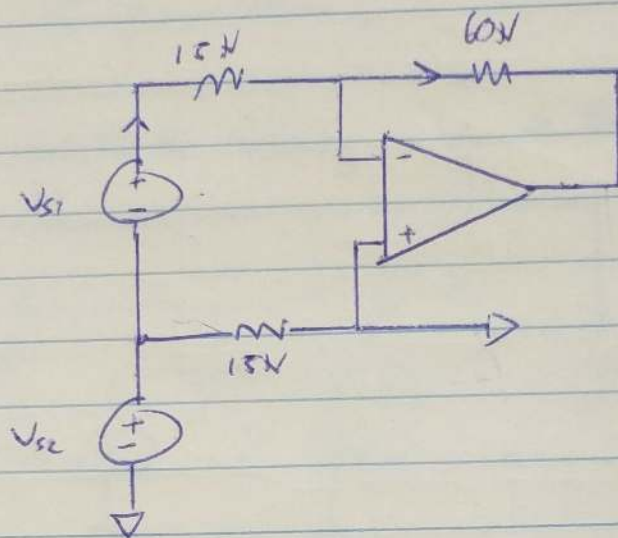
$$\frac{V_{s1}}{15k} + \frac{V_{s2}}{15k} - \frac{4V_{s2}}{75k} - \frac{4V_{s2}}{300k} = \frac{-V_o}{60k}$$

$$\frac{V_{s1}}{15k} = -\frac{V_o}{60k}$$

$$\Rightarrow V_o = -\frac{60k}{15k} V_{s1} = \underline{\underline{-4V_{s1}}}$$

(2)

Now, let us repeat the analysis with the switch closed



Now $V_+ = V_- = 0\text{V}$, and

$$\frac{V_{S1} + V_{S2}}{15\text{ k}} = \frac{-V_o}{60\text{ k}}$$

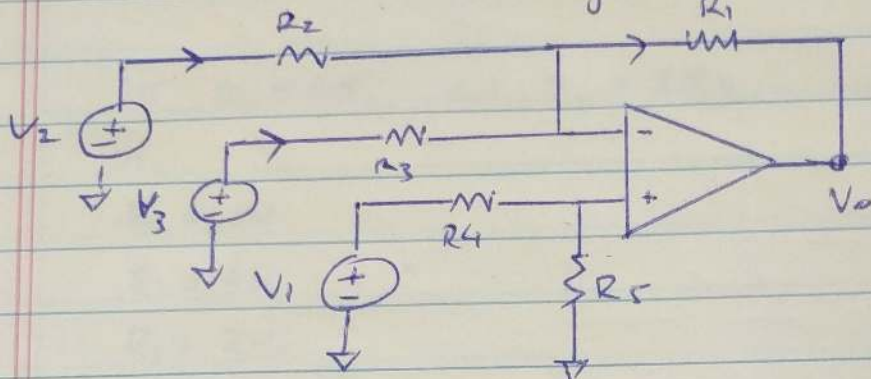
$$\Rightarrow V_o = \underline{\underline{-4(V_{S1} + V_{S2})}}$$

(3)

4.45

One possible solution using a single op-amp is shown below.

We want to realize the following: $V_o = 5V_1 - 4V_2 - 2V_3$



Let's use superposition to solve for the output of the circuit.

$$V_1 = V_3 = 0,$$

$$V_o = -\frac{R_1}{R_2} V_2$$

$$V_1 = V_2 = 0$$

$$V_o = -\frac{R_1}{R_3} V_3$$

$$V_2 = V_3 = 0$$

$$V_+ = \frac{V_1}{R_4 + R_5}, \text{ and}$$

$$-\frac{V_+}{R_2 \parallel R_3} = \frac{V_+ - V_o}{R_1}$$

$$\frac{V_o}{R_1} = \frac{V_+}{R_1} + \frac{V_+}{R_2 \parallel R_3}$$

$$V_o = V_+ R_1 \cdot \frac{(R_1 + R_2 \parallel R_3)}{R_1 (R_2 \parallel R_3)} = \frac{V_1 R_5}{R_4 + R_5} \cdot \frac{(R_1 + R_2 \parallel R_3)}{R_2 \parallel R_3}$$

(4)

Summary gives

$$V_o = -\frac{R_1}{R_2} V_2 - \frac{R_1}{R_3} V_3 + \frac{V_1 R_5}{R_4 + R_5} \left(\frac{R_1 + R_2 \parallel R_3}{R_2 \parallel R_3} \right)$$

let $R_1 = 4R_2$, and $R_1 = 2R_3$.

s.t.

$$R_2 = 1k$$

$$R_1 = 4k$$

$$R_3 = 2k,$$

now : $V_o = -4V_2 - 2V_3 + \frac{V_1 R_5}{R_4 + R_5} \cdot 7$

so: $\frac{7R_5}{R_4 + R_5} = 5$, let $R_5 = 1k$

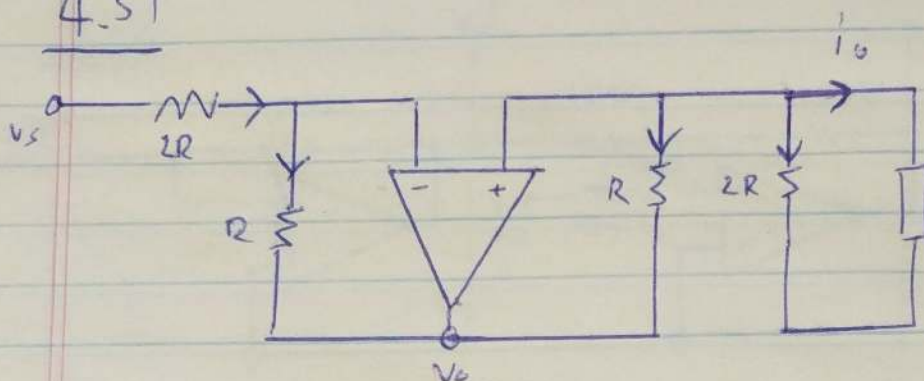
s.t. $7000 = 5R_4 + 5000$

$$\underline{\underline{R_4 = 400}}$$

which achieves an objective

(5)

4.51



$$i_o = -\frac{V_+}{2R} - \frac{(V_+ - V_o)}{R} \Rightarrow i_o = -\frac{V_+}{2R} - \frac{V_+}{R} + \frac{V_o}{R}$$

$$i_o = -V_+ \left(\frac{3}{2R} \right) + \frac{V_o}{R}$$

Now, let's do a nodal analysis at the negative terminal

$$\frac{V_s - V_+}{2R} = \frac{V_+ - V_o}{R}$$

$$\Rightarrow \frac{V_s}{2R} + \frac{V_o}{R} = \frac{V_+}{R} + \frac{V_+}{2R}$$

$$\Rightarrow \frac{V_s}{2R} + \frac{V_o}{R} = \frac{3}{2R} V_+$$

$$V_+ = \left(\frac{V_s}{2R} + \frac{V_o}{R} \right) \frac{2R}{3}$$

$$V_+ = \frac{RV_s + 2RV_o}{3R}$$

now combine the equations

$$i_o = -\frac{(RV_s + 2RV_o) \cdot \frac{3}{2R}}{3R} + \frac{V_o}{R}$$

$$= -\frac{RV_s}{2R^2} - \frac{2RV_o \cdot 3}{3 \cdot 2 \cdot R^2} + \frac{V_o}{R}$$

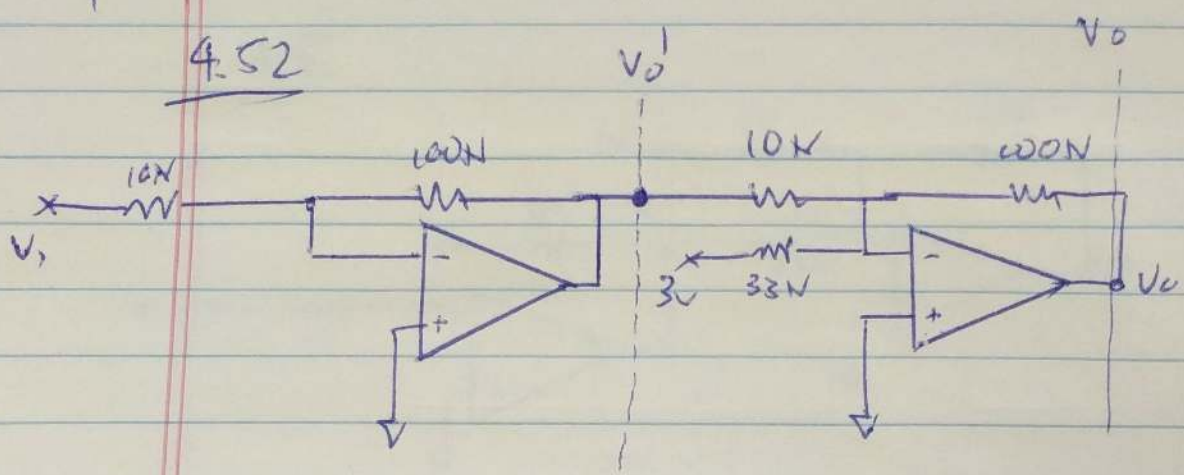
$$= -\frac{V_s}{2R}$$

(6)

Inverting opamp -

Inverting Summer

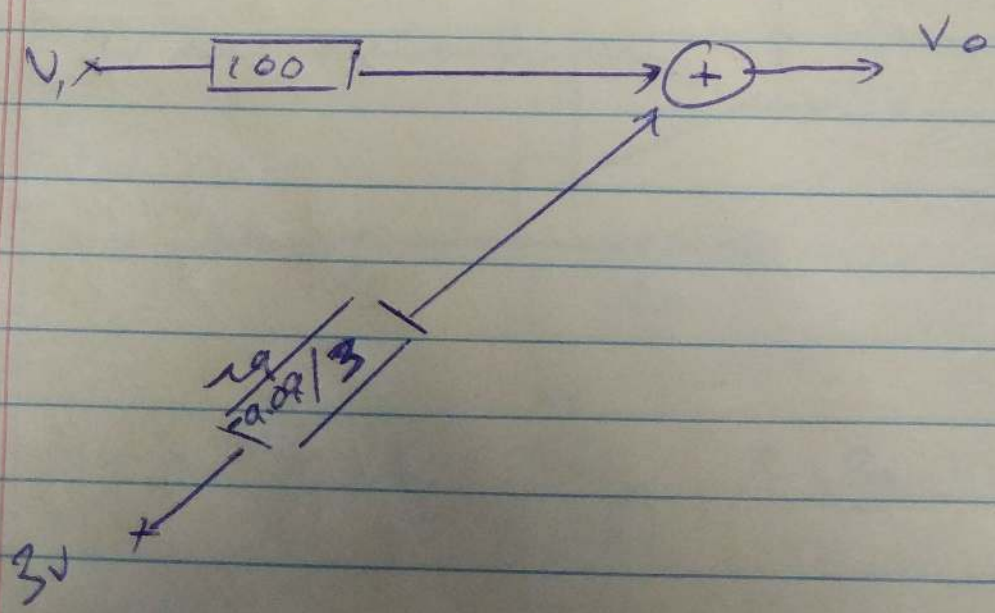
4.52



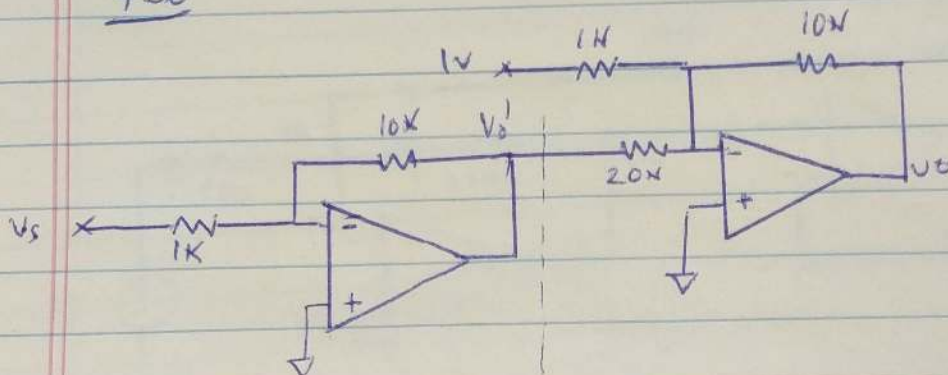
$$V_0' = -10V_1$$

$$V_0 = -10V_0' - \frac{100k}{33k} \cdot 3V$$

$$= 100V_1 - 9.09$$



4.56

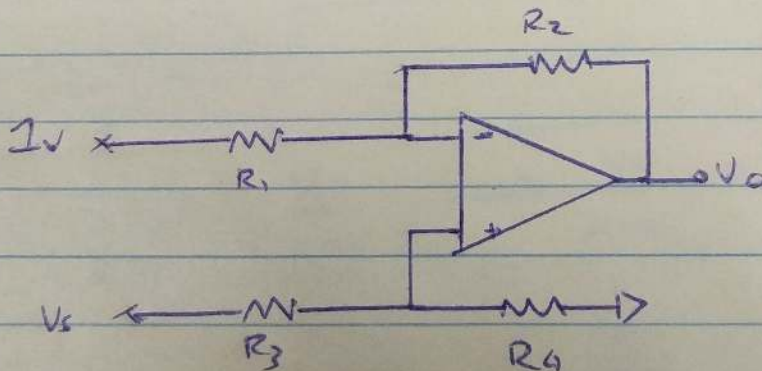


$$V_o' = -10V_s$$

$$V_o = -\frac{1}{2} V_o' - \frac{10 \cdot 1}{1}$$

$$V_o = \underline{\underline{5V_s - 10}}$$

Now, we can redesign using a single opamp.



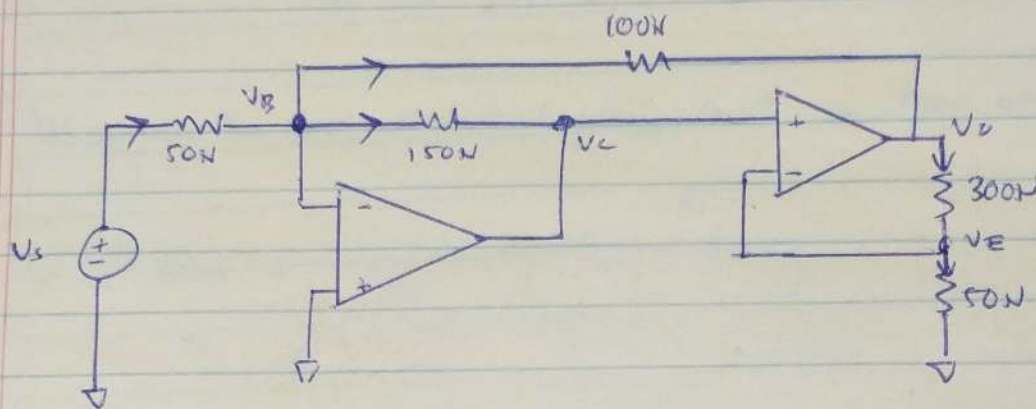
$$V_o = \frac{-R_2}{R_1} \cdot 1 + \frac{R_4}{R_3 + R_4} \cdot \frac{R_1 + R_2}{R_1} \cdot V_s \quad \left(\text{where we} \right.$$

can follow the same steps as before to get the)

set $R_2 = 10K$ $R_4 = 5K$

$R_1 = 1K$ $R_3 = 6K$ to achieve desired function

4-60



Nodal analysis at V_B

$$\frac{V_s}{50k} = -\frac{V_O}{100k} - \frac{V_C}{150k} \Rightarrow V_O = -100k \left(\frac{V_C}{150k} + \frac{V_s}{75k} \right)$$

Nodal analysis at V_E

$$\frac{V_O - V_C}{300k} = \frac{V_E}{50k} \Rightarrow \frac{V_O}{300k} = \frac{V_C}{300k} + \frac{V_C}{50k}$$

$$\frac{V_O}{300k} = \frac{7V_C}{300k} \Rightarrow \underline{V_C = \frac{V_O}{7}}$$

Sub into first equation

$$V_O = -100k \left(\frac{V_O}{7 \cdot 150k} + \frac{V_s}{75k} \right)$$

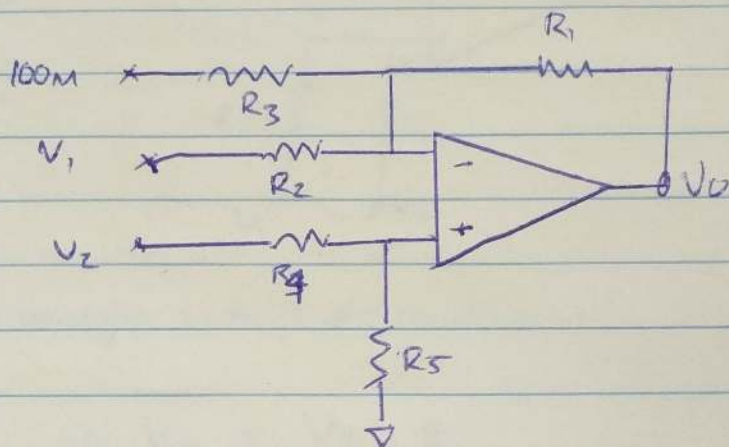
$$V_O \left(1 + \frac{100k}{7 \cdot 150k} \right) = -\frac{2}{7} V_s$$

$$V_O \left(\frac{23}{21} \right) = -\frac{2}{7} V_s \Rightarrow \underline{V_O = -\frac{42}{23} V_s}$$

4.66

$$V_o = -3V_1 + 2V_2 - 300\text{mV}$$

We can use the same circuit topology that we have used before



$$V_o = -\frac{R_1}{R_3} \cdot 100\text{m} - \frac{V_1 R_1}{R_2} + \frac{V_2 R_5}{R_4 + R_5} \cdot \frac{R_1 + R_2 \parallel R_3}{R_2 \parallel R_3}$$

let. $R_1 = 6000$, $R_2 = 3000$, $R_3 = 2000$

then: $\frac{11}{6} \frac{V_2 R_5}{R_4 + R_5} = \frac{11}{6} 2V_2 \Rightarrow \frac{R_5}{R_4 + R_5} = \frac{2}{6}$

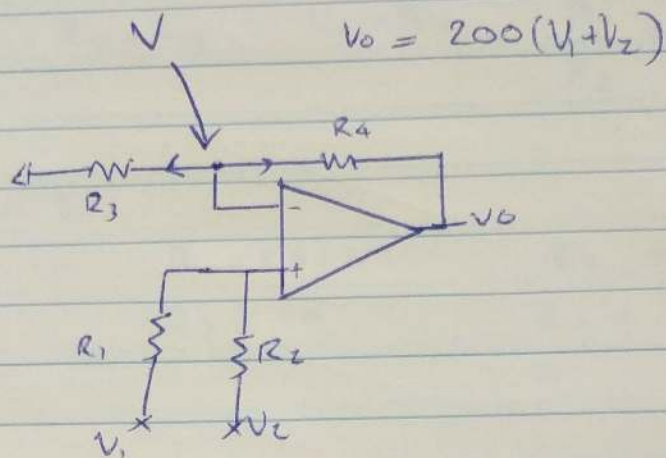
let $\frac{R_5}{R_4 + R_5} = \frac{2}{6} \Rightarrow \frac{R_5}{R_4 + R_5} = \frac{1}{3}$

$$\frac{R_5}{R_4 + R_5} \cdot 6 = 2 \Rightarrow \frac{R_5}{R_4 + R_5} = \frac{2}{6} = \frac{1}{3}$$

so,

$$\boxed{\begin{matrix} R_5 = 1\text{K} \\ R_4 = 2\text{K} \end{matrix}}$$

467



We can analyze with super position.

$$V_1 = 0 \Rightarrow V_+ = \frac{V_2 R_1}{R_1 + R_2}$$

Now, let's apply nodal analysis at negative terminal ($V_- = V_0$)

$$\frac{V}{R_3} + \frac{V - V_0}{R_4} = 0 \Rightarrow \frac{V_0}{R_4} = V \left(\frac{1}{R_3} + \frac{1}{R_4} \right)$$

$$V_0 = VR_4 \left(\frac{R_3 + R_4}{R_3 R_4} \right)$$

$$= V \left(\frac{R_3 + R_4}{R_3} \right)$$

$$V_0 = \frac{V_2 R_1 (R_3 + R_4)}{R_1 + R_2 R_3}$$

Applying the same procedure with $V_2 = 0$, gives

$$V_0 = \frac{V_2 R_1}{R_1 + R_2} \cdot \frac{R_3 + R_4}{R_3} + \frac{V_1 R_2}{R_1 + R_2} \cdot \frac{(R_3 + R_4)}{R_3}$$

Now let, $R_3 + R_4 = R_1 + R_2$, $\frac{R_1}{R_3} = 200$, $\frac{R_2}{R_3} = 200$

For instance :

$$R_1 = 200 \text{ N}$$

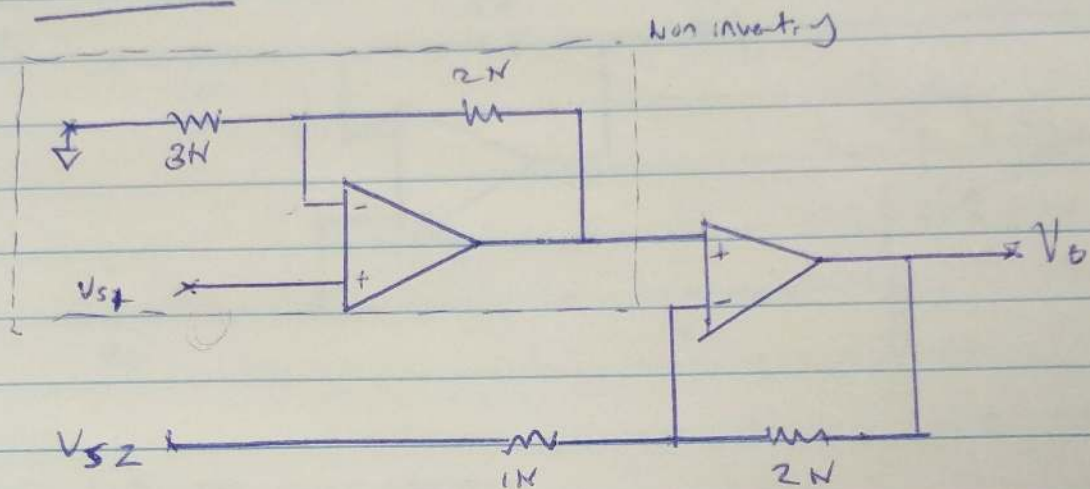
$$R_2 = 200 \text{ N}$$

$$R_3 = 1 \text{ H}$$

$$R_4 = 399 \text{ K}$$

4.25

Circuit #1



Let's use super position.

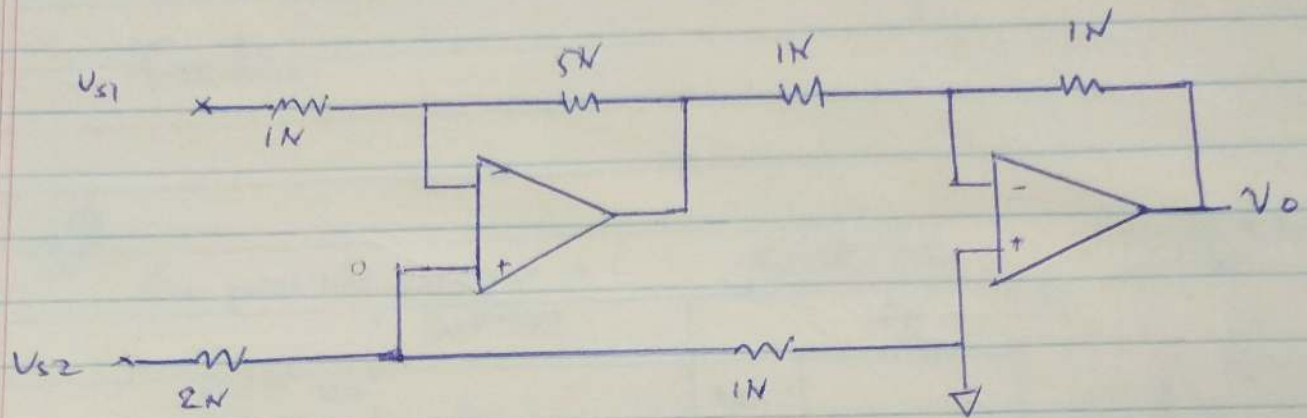
$$V_{s2} = 0 \Rightarrow V_o = \left(1 + \frac{2}{3}\right) \left(1 + \frac{2}{1}\right) V_{s1}$$

$$= 5V_{s1}$$

$$V_{s1} = 0 \Rightarrow -2V_{s2}$$

$$\Rightarrow V_o = 5V_{s1} - 2V_{s2}$$

Circuit 2/



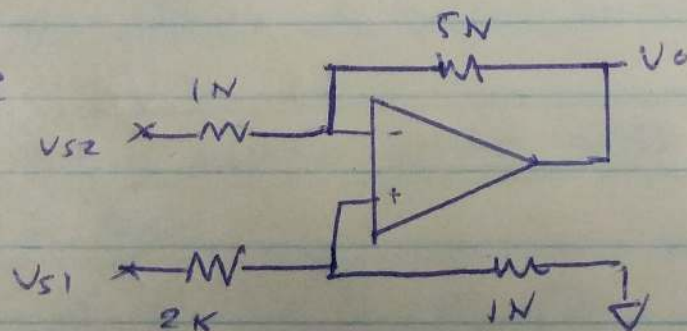
Again, let's use superposition

$$V_{s2} = 0 \Rightarrow V_o = (-5)(-1)V_{s1}$$

$$V_{s1} = 0 \Rightarrow \left(\frac{V_{s2}}{3}\right)(6)(-1)$$

$$\text{So, } V_o = 5V_{s1} - 2V_{s2}$$

Circuit 3



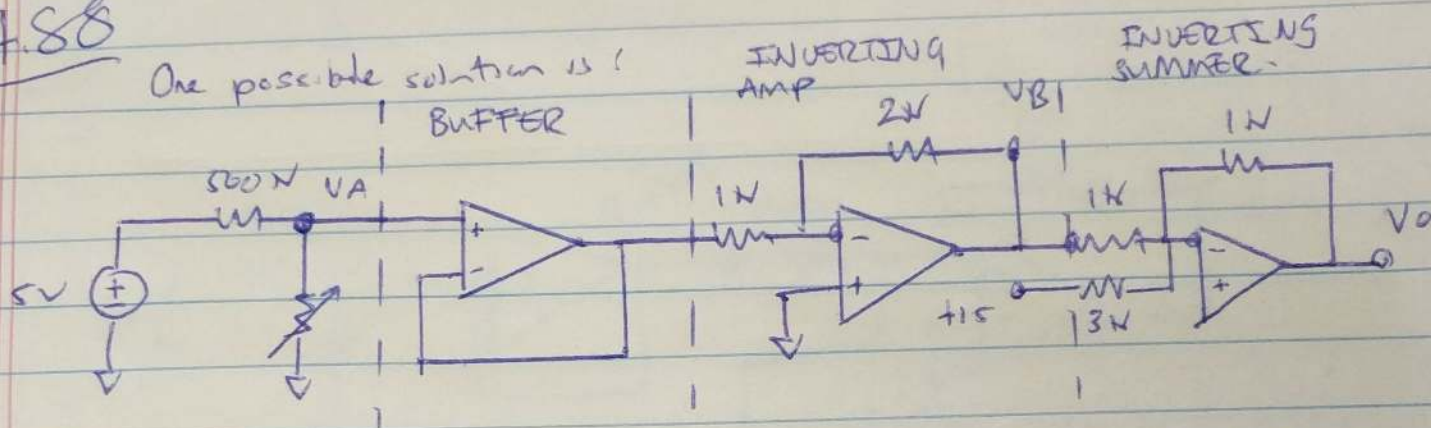
$$V_{s1} = 0 \Rightarrow V_o = -5V_{s2}$$

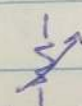
} so we see already that this circuit will not work

Circuit #1 is best, has least number of resistors and is able to perform required function.

4.88

One possible solution is:



Here  represents the photoresistor. When dark, the output V_A is 2.5V, when bright it's 9.9998×10^{-5} Volts.

• Voltage buffer so that the amplifier gains are not affected by the photoresistor voltage divider circuit

• Next is inverting amplifier circuit with a gain of **2**. If dark, the $V_B = -2.5$, if bright, then 1.99996×10^{-4} .

• Next is inverting summer which adds $-5V$ to output. So, if dark then

$$V_O = -(-5) - 5 = 0V$$

If bright

$$V_O = -1.99996 \times 10^{-4} - 5 \approx \underline{\underline{-5 \text{ Volts}}}$$