

MAE 143b

Spring 2014

HW #6

4.2c

$$G = \frac{a}{s+a} \quad \text{and} \quad K = k_1 \frac{s+c}{s+b}$$

$$S = \frac{1}{1+GK} = \frac{(s+a)(s+b)}{(s+a)(s+b) + ak_1(s+c)} \quad \text{second order}$$

$$D = \frac{G}{1+GK} = \frac{a(s+b)}{(s+a)(s+b) + ak_1(s+c)} \quad \text{Second order}$$

$$H = \frac{GK}{1+GK} = \frac{ak_1(s+c)}{(s+a)(s+b) + ak_1(s+c)} \quad \text{Second order}$$

$$Q = \frac{k}{1+GK} = \frac{k_1(s+a)(s+c)}{(s+a)(s+b) + ak_1(s+c)} \quad \text{second order}$$

4.2e

$$G = \frac{a_2}{s^2 + a_1 s + a_2} \quad \text{and} \quad K = k_p + k_d s$$

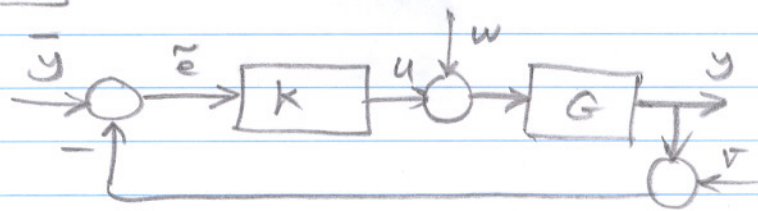
$$S = \frac{1}{1 + GK} = \frac{s^2 + a_1 s + a_2}{s^2 + a_1 s + a_2 + (k_p + k_d s) a_2} \quad \text{second order}$$

$$D = \frac{G}{1 + GK} = \frac{a_2}{s^2 + a_1 s + a_2 + (k_p + k_d s) a_2} \quad \text{second order}$$

$$H = \frac{GK}{1 + GK} = \frac{a_2 (k_p + k_d s)}{s^2 + a_1 s + a_2 + (k_p + k_d s) a_2} \quad \text{second order}$$

$$Q = \frac{K}{1 + GK} = \frac{(k_p + k_d s)(s^2 + a_1 s + a_2)}{s^2 + a_1 s + a_2 + (k_p + k_d s) a_2} \quad \text{third order}$$

4.7



$$\omega = 0$$

$$G = \frac{1}{s+1} \quad \text{and} \quad K = \frac{1}{s}$$

$$\bar{y}(t) = \bar{y} \quad v(t) = \bar{v} + \cos(\omega t) \quad t \geq 0$$

Since we have $\cos(\omega t)$ function as one of input, thus we need to use below equation to find the steady-state response. Also this can also be used for a constant input considering $\omega = 0$. Here is the steady-state response of system considering all inputs.

$$y_{ss}(t) = \bar{y} S(0) - \bar{v} S(0) + |S(j\omega)| \cos(\omega t + \angle S(j\omega))$$

$S(0) = \lim_{\omega \rightarrow 0} S(j\omega)$ and $S(j\omega)$ is the transfer function.

finding $S(j\omega)$: $Y(s) = G U(s) = G K \tilde{E}(s) = G K (\bar{Y}(s) - V(s) - Y(s))$

$$\Rightarrow S(s) = \frac{GK}{1 + GK} = \frac{1}{s(s+1) + 1} \quad \begin{matrix} s=j\omega \\ \hline = \frac{1}{j\omega + (1-\omega^2)} \end{matrix}$$

$$S(0) = 1$$

$$|S(j\omega)| = \frac{1}{\sqrt{\omega^2 + (1-\omega^2)^2}} \quad \angle S(j\omega) = -\arctan\left(\frac{\omega}{1-\omega^2}\right)$$

$$\Rightarrow y_{ss}(t) = \bar{y} - \bar{v} + \frac{1}{\sqrt{\omega^2 + (1-\omega^2)^2}} \cos(\omega t - \arctan(\frac{\omega}{1-\omega^2}))$$

$$e_{ss}(t) = \bar{y} - y_{ss}(t) = \bar{v} - \frac{1}{\sqrt{\omega^2 + (1-\omega^2)^2}} \cos(\omega t - \arctan(\frac{\omega}{1-\omega^2}))$$

→ The closed-loop does not achieve asymptotic tracking of the reference input ($\bar{y}(t)$) because there are non-zero terms in steady-state response.

→ NO, the closed-loop does not achieve asymptotic rejection of the measurement noise input.

→ If $\bar{v} = 0$ and $W \gg 0 \Rightarrow$ tracking is achieved because in this case $y_{ss}(t) = \bar{y}$.

4.13

$$J\ddot{\omega} + (b_1 + b_2)\omega = \tau + gr(m_1 - m_2)$$

$$J = J_1 + J_2 + r^2(m_1 + m_2)$$

$$\tau_1 = r\omega \Rightarrow \omega = \frac{\tau_1}{r}$$

$$r = 1\text{ m}$$

$$m_1 = m_2 = 1000\text{ kg}$$

$$b_1 = b_2 = 120\text{ kg m/s}$$

$$J_1 = J_2 = 200\text{ kg m}^2$$

$$g = 10\text{ m/s}^2$$

$$\tau = k(\bar{x}_1 - x_1), x_1 = \int_0^t \dot{x}_1(\tau) d\tau \Rightarrow \ddot{x}_1 = \dot{v}_1(t) \quad (**)$$

$$\text{Replacing } \omega \text{ with } v_1 \Rightarrow J\dot{v}_1 + (b_1 + b_2)v_1 = \tau \cdot r + gr^2(m_1 - m_2)$$

$$\text{Replacing } v_1 \text{ with } x_1 \Rightarrow J\ddot{x}_1 + (b_1 + b_2)\dot{x}_1 = \tau \cdot r + gr^2(m_1 - m_2)$$

$$\text{Replacing } \tau = k(\bar{x}_1 - x_1) \Rightarrow J\ddot{x}_1 + (b_1 + b_2)\dot{x}_1 = k(\bar{x}_1 - x_1) \cdot r + gr^2(m_1 - m_2)$$

Taking Laplace from both sides of equation:

$$J s^2 X_1(s) + (b_1 + b_2) s X_1(s) = k(\bar{X}_1(s) - X_1(s)) \cdot r + \frac{gr^2}{s}(m_1 - m_2)$$

$$\Rightarrow X_1(s) \left(J s^2 + (b_1 + b_2) s + kr \right) = k \bar{X}_1(s) r + \frac{gr^2}{s}(m_1 - m_2)$$

$$\Rightarrow X_1(s) = \frac{r k \bar{X}_1(s)}{J s^2 + (b_1 + b_2) s + kr} + \frac{gr^2(m_1 - m_2)}{s(J s^2 + (b_1 + b_2) s + kr)}$$

Roots for denominator:

$$J s^2 + (b_1 + b_2) s + K r = 0 \Rightarrow s = \frac{-(b_1 + b_2) \pm \sqrt{(b_1 + b_2)^2 - 4 J K r}}{2 J}$$

Inserting the given values:

$$\Rightarrow s = \frac{-240 \pm \sqrt{(240)^2 - 4(2400) \times K}}{2 \times 2400}$$

In order to make the system internally stable:

$$s = \frac{-240 \pm \sqrt{57600 - 9600K}}{4800} < 0$$

$$\Rightarrow \boxed{K > 0}$$

$$\left. \begin{array}{l} \lim_{t \rightarrow \infty} x_1(t) = \lim_{s \rightarrow 0} s X_1(s) \\ \bar{X}_1(s) = \frac{\bar{x}_1}{s} \\ m_1 = m_2 \end{array} \right\} \Rightarrow \lim_{s \rightarrow 0} s X_1(s) = \bar{x}_1$$

Yes. As shown, the closed-loop system asymptotically track a constant position reference.

4.14

From answer to 4.13, we have;

$$X_1(s) = \frac{rk\bar{x}_1(s)}{Js^2 + (b_1 + b_2)s + kr} + \frac{gr^2(m_1 - m_2)}{s(Js^2 + (b_1 + b_2)s + kr)}$$

$$Js^2 + (b_1 + b_2)s + kr = 0$$

$$\Rightarrow s = \frac{-(b_1 + b_2) \pm \sqrt{(b_1 + b_2)^2 - 4Jkr}}{2J}$$

Inserting the given values:

$$s = \frac{-240 \pm \sqrt{57600 - 4(2200)k}}{2 \times 2200}$$

$$\Rightarrow s = \frac{-240 \pm \sqrt{57600 - 8800k}}{4400}$$

Again, $k > 0$ so the system is internally stable.

$$\left. \begin{array}{l} \lim_{t \rightarrow \infty} x_1(t) = \lim_{s \rightarrow 0} sX_1(s) \\ \bar{x}_1(s) = \frac{x_1}{s} \\ m_1 = 1000 \text{ kg} \\ m_2 = 800 \text{ kg} \end{array} \right\} \Rightarrow \lim_{s \rightarrow 0} sX_1(s) = \bar{x}_1 + \frac{2000}{k}$$

No. As shown, the closed-loop system cannot track a constant position.